



Efficient and Accurate Computation of Eye Diagrams and Bit Error Rates in a Single-channel CRZ System

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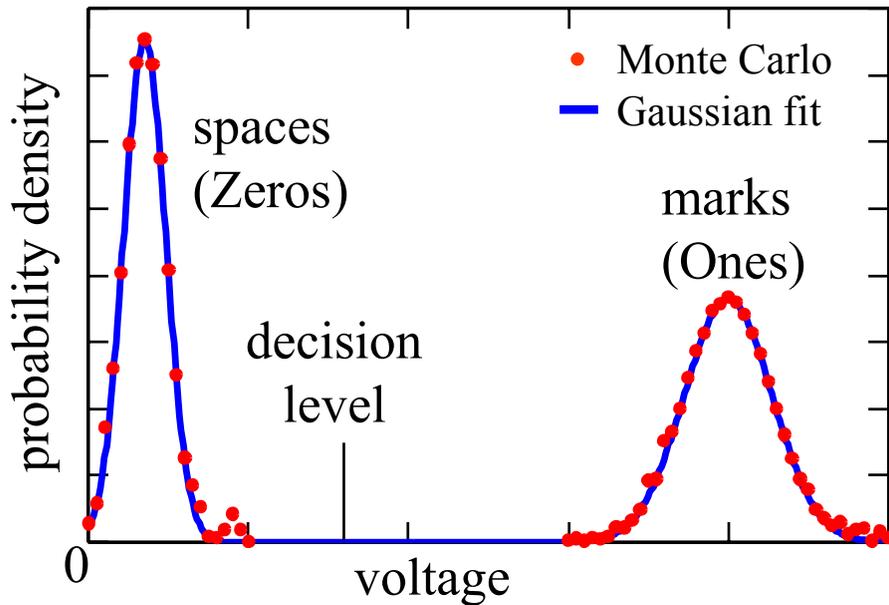
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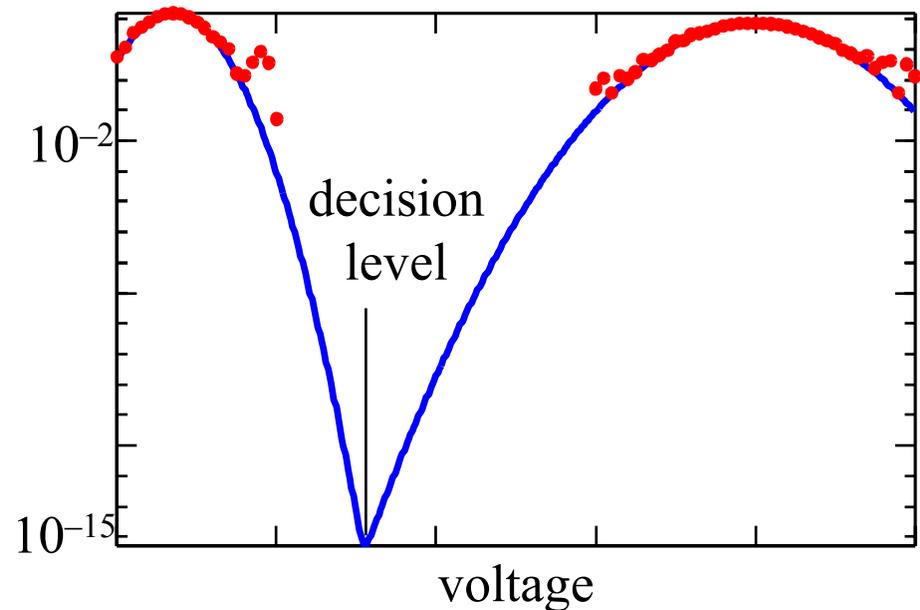
Calculating Bit Error Rates

BERs are commonly calculated from Monte Carlo simulation results using Gaussian extrapolation

Linear scale



Log scale



But: voltage probability densities are not Gaussian [Marcuse 1990]



Outline

Goal

Accurate calculation of BER vs. decision level

- ① Linearize noise propagation
- ② Include signal-noise beating and data modulation
- ③ Model a realistic electrical receiver filter

Approach

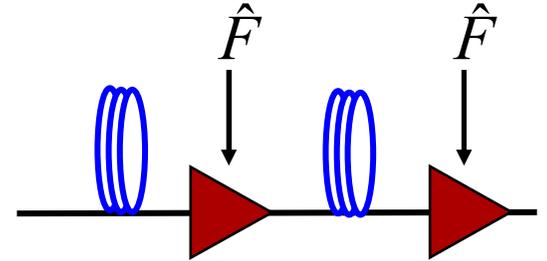
Calculate the multivariate Gaussian noise pdf of the optical field
Justification: Noise-noise interaction in the fiber is small

Linearizing the NLS

Nonlinear Schrödinger equation with ASE noise

$$i \frac{\partial u}{\partial z} + \frac{D(z)}{2} \frac{\partial^2 u}{\partial t^2} + |u|^2 u = ig(z)u + \hat{F}$$

\hat{F} : added Gaussian white noise



Now set $u = u_0 + \delta u$, $u_0 = \langle u \rangle$: noise-free signal

δu : accumulated noise

$$i \frac{\partial \delta u}{\partial z} + \frac{D(z)}{2} \frac{\partial^2 \delta u}{\partial t^2} + 2|u_0|^2 \delta u + u_0^2 (\delta u)^* = ig(z)\delta u + \hat{F}$$

Doob's Theorem: δu is multivariate Gaussian distributed

Noise Covariance Matrix

$$\delta u(t) = \sum_{k=-N_{FFT}/2}^{N_{FFT}/2-1} [\alpha_k + i\beta_k] \exp(i\omega_k t), \quad N_{FFT} = 2048$$

$$\mathbf{a} = (\alpha_{-N/2}, \dots, \alpha_{N/2-1}, \beta_{-N/2}, \dots, \beta_{N/2-1})^T, \quad N = 80$$

Covariance matrix $\mathbf{K}_{kl} = \langle \mathbf{a}_k \mathbf{a}_l \rangle$, $\mathbf{K} = \langle \mathbf{a} \mathbf{a}^T \rangle$

Multivariate Gaussian distribution of \mathbf{a} :

$$f_{\mathbf{a}}(\mathbf{a}, z) = (2\pi)^{-2N} \sqrt{\det \mathbf{K}^{-1}} \exp\left(-\frac{1}{2} \mathbf{a}^T \mathbf{K}^{-1} \mathbf{a}\right)$$

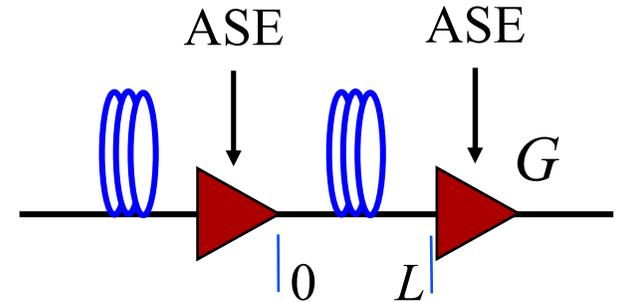
How to Compute the Covariance Matrix

Solve the linearized homogeneous propagation equation

$$\frac{\partial \mathbf{a}}{\partial z} = \mathbf{R}(z) \mathbf{a} \Rightarrow \mathbf{a}(L) = \mathbf{\Psi} \mathbf{a}(0)$$

$$\mathbf{K}(L) = G \mathbf{\Psi} \mathbf{K}(0) \mathbf{\Psi}^T + \eta \mathbf{I}$$

↑ ASE input



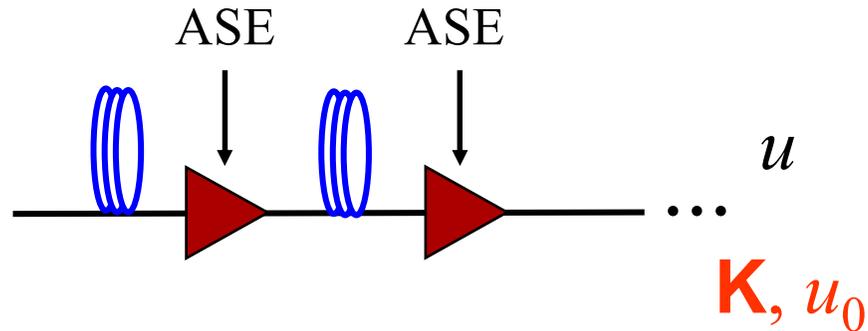
But: ODE is stiff due to dispersion term. Solution: perturbative approach

$$\mathbf{a}^{(k)}(0) = \varepsilon \hat{e}_k \xrightarrow{\text{fiber}} \mathbf{a}^{(k)}(L) \Rightarrow \mathbf{\Psi}_{jk} = \frac{\mathbf{a}_j^{(k)}(L)}{\varepsilon}$$

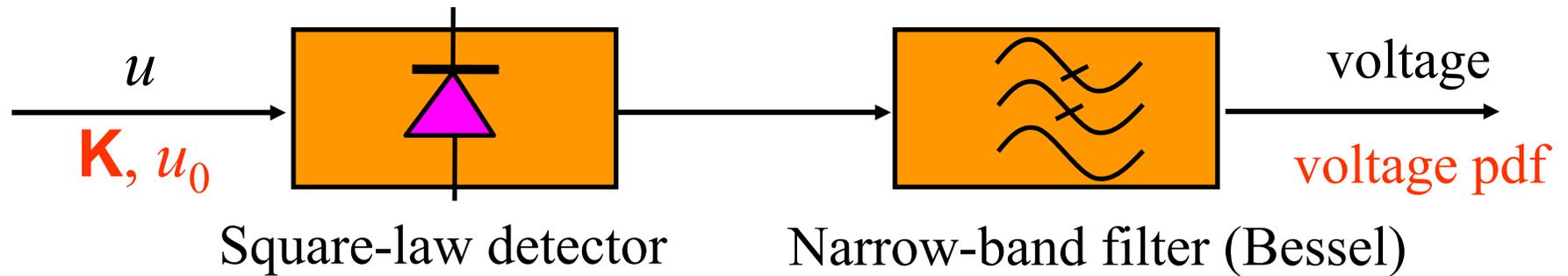
Compute $\mathbf{\Psi}$ by perturbing each of the N frequency modes separately

Simulation Setup

Propagation:



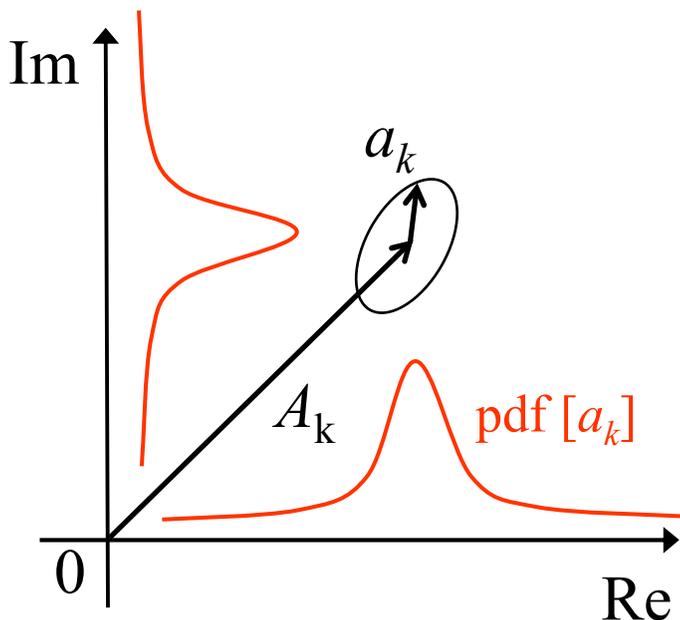
Receiver model:



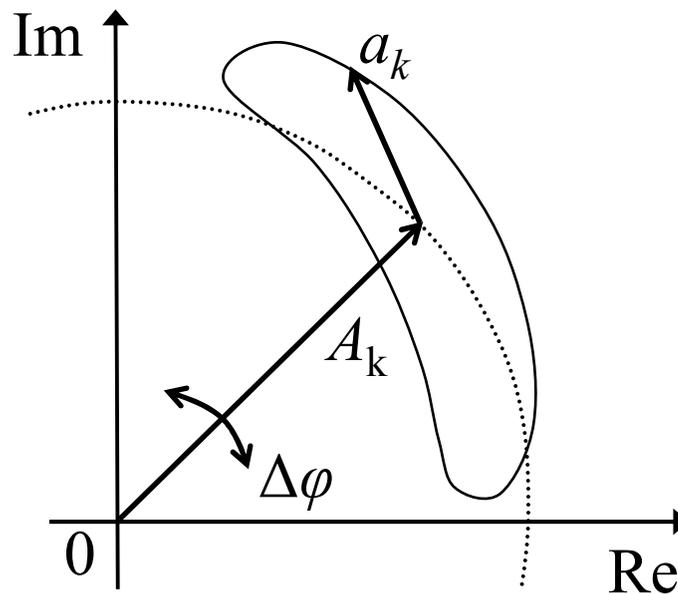
Quadratic noise-noise terms in the receiver cannot be neglected !

Strong Jitter Distorts the Gaussian pdf

Small phase jitter



Large phase jitter



Separate phase jitter from $\mathbf{a}^{(k)}(L)$

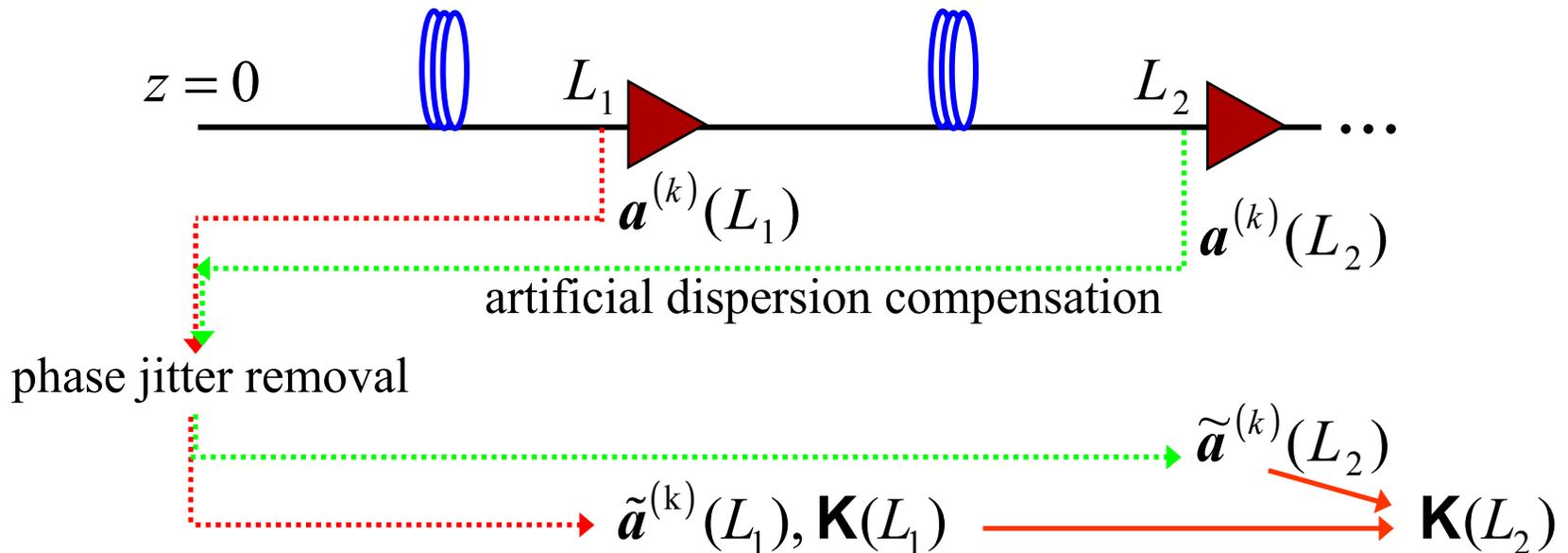
Phase jitter rotates signal around origin, distorting the Gaussian pdf

Phase Jitter Removal

Remove phase jitter by projecting out the noise proportional to iu_0

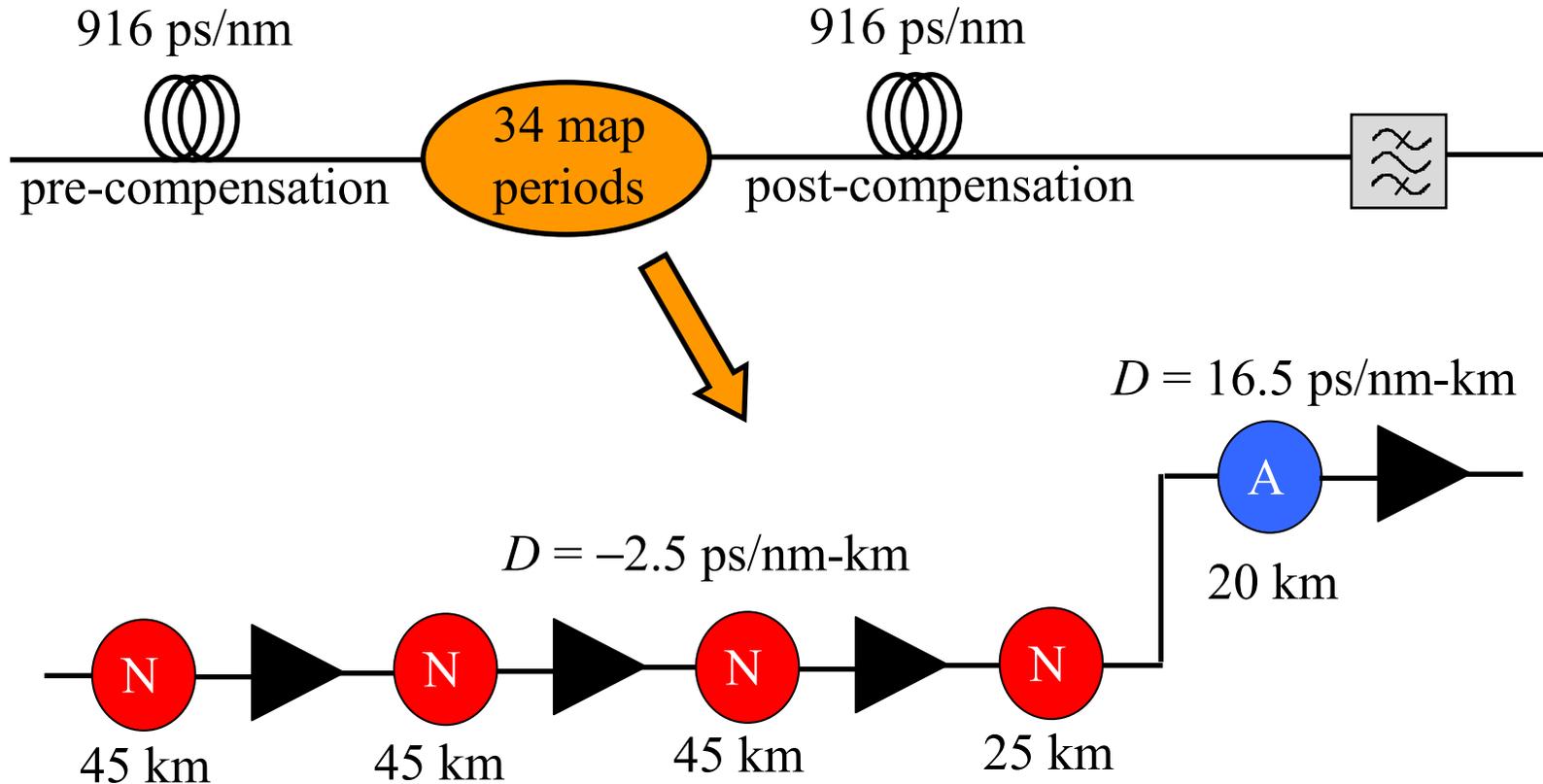
$$\tilde{\mathbf{a}}^{(k)} = \mathbf{a}^{(k)} - \frac{(\mathbf{a}^{(k)}, \mathbf{v})}{(\mathbf{v}, \mathbf{v})} \mathbf{v}, \quad \mathbf{v} = \text{FT}\{iu_0(t)\}$$

Jitter removal requires artificial dispersion compensation in CRZ:



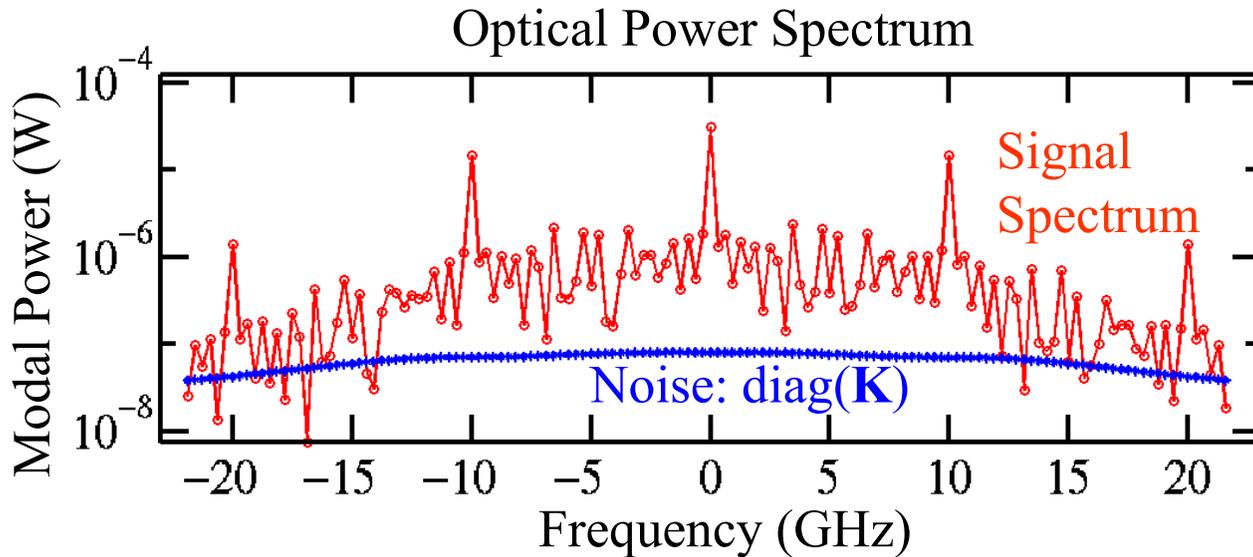
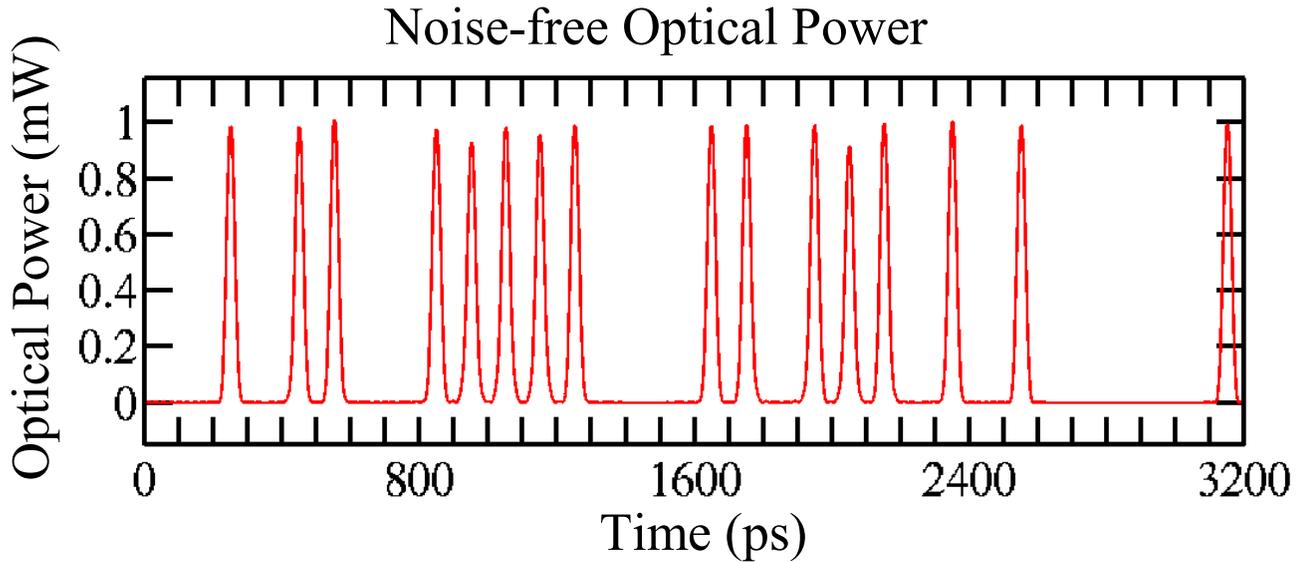
Test System: Submarine CRZ, 6100 km

Modeled on a transatlantic communications system by Tyco Communications, Inc.

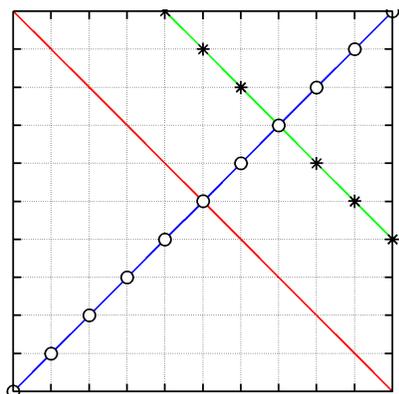


Non-periodic evolution: medium nonlinearity, but strong pulse overlap

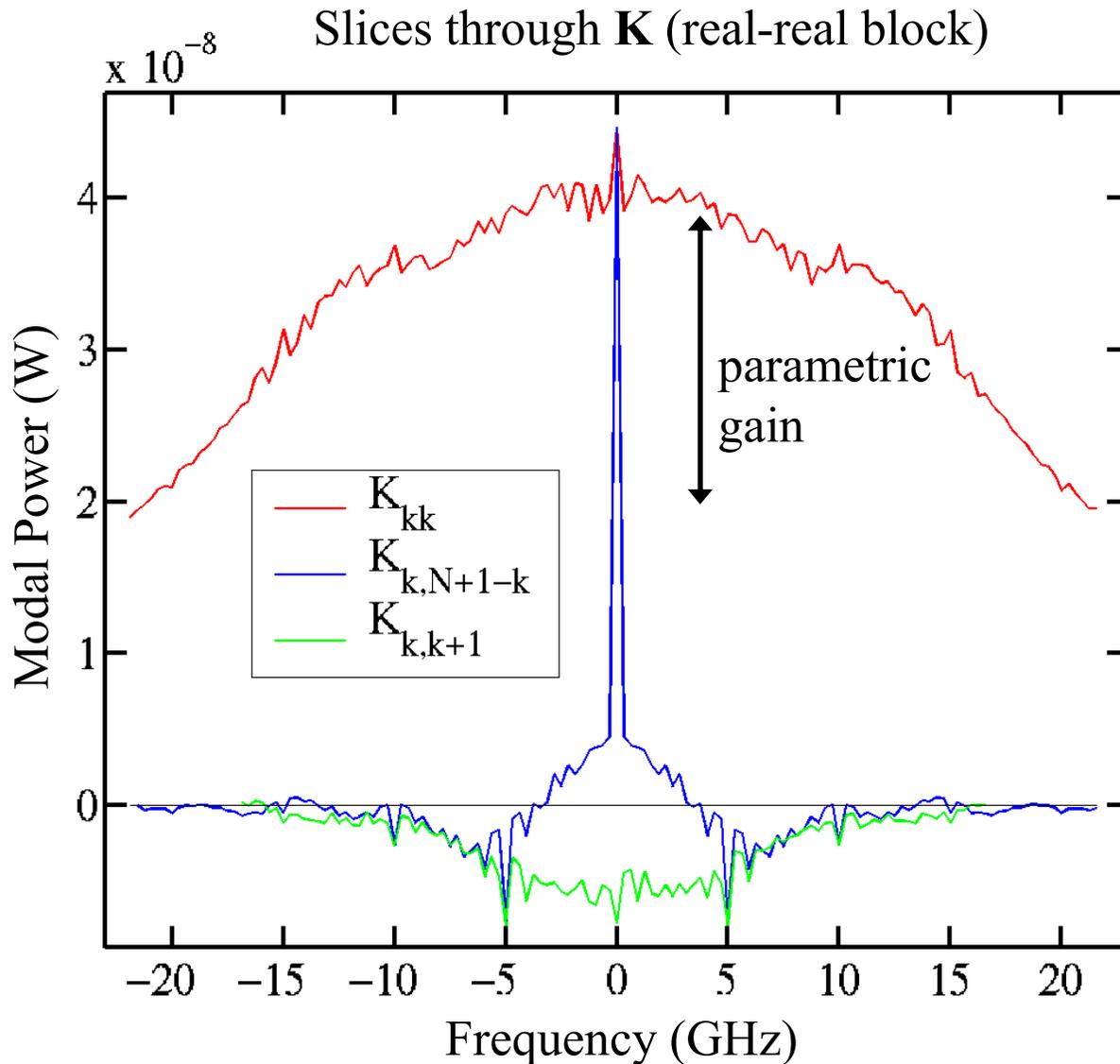
Noise-Free Optical Signal at Receiver



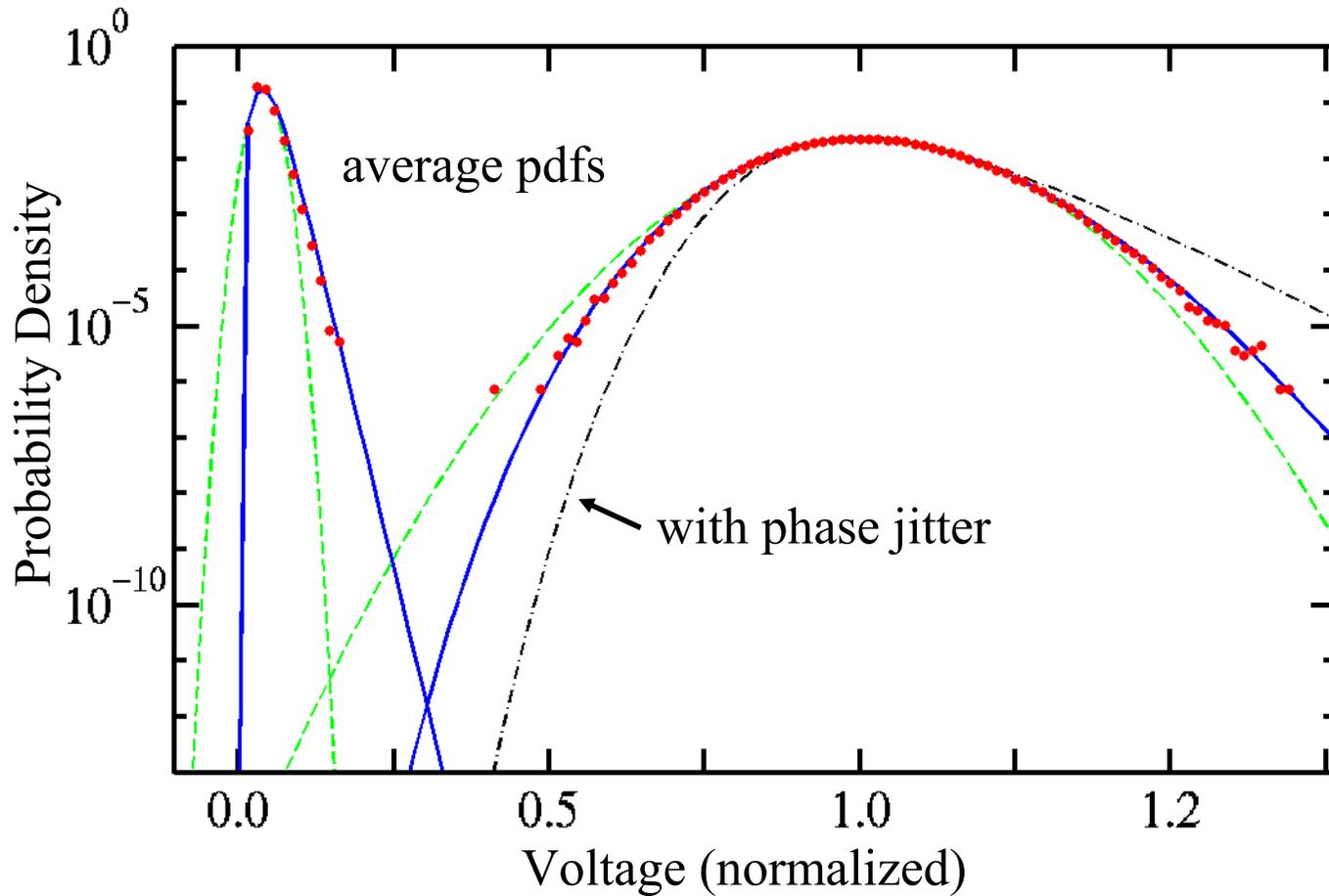
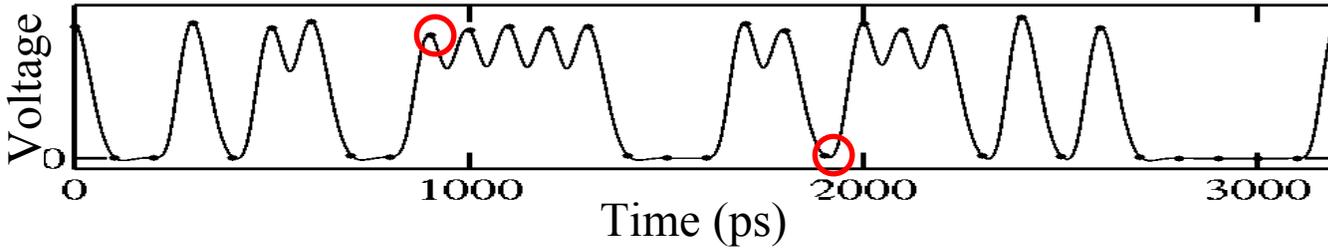
Characterizing the Covariance Matrix



Matrix sketch

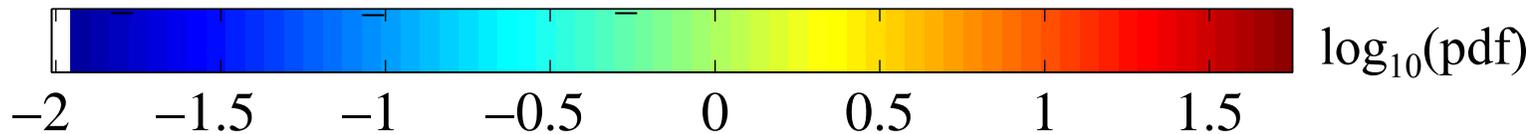
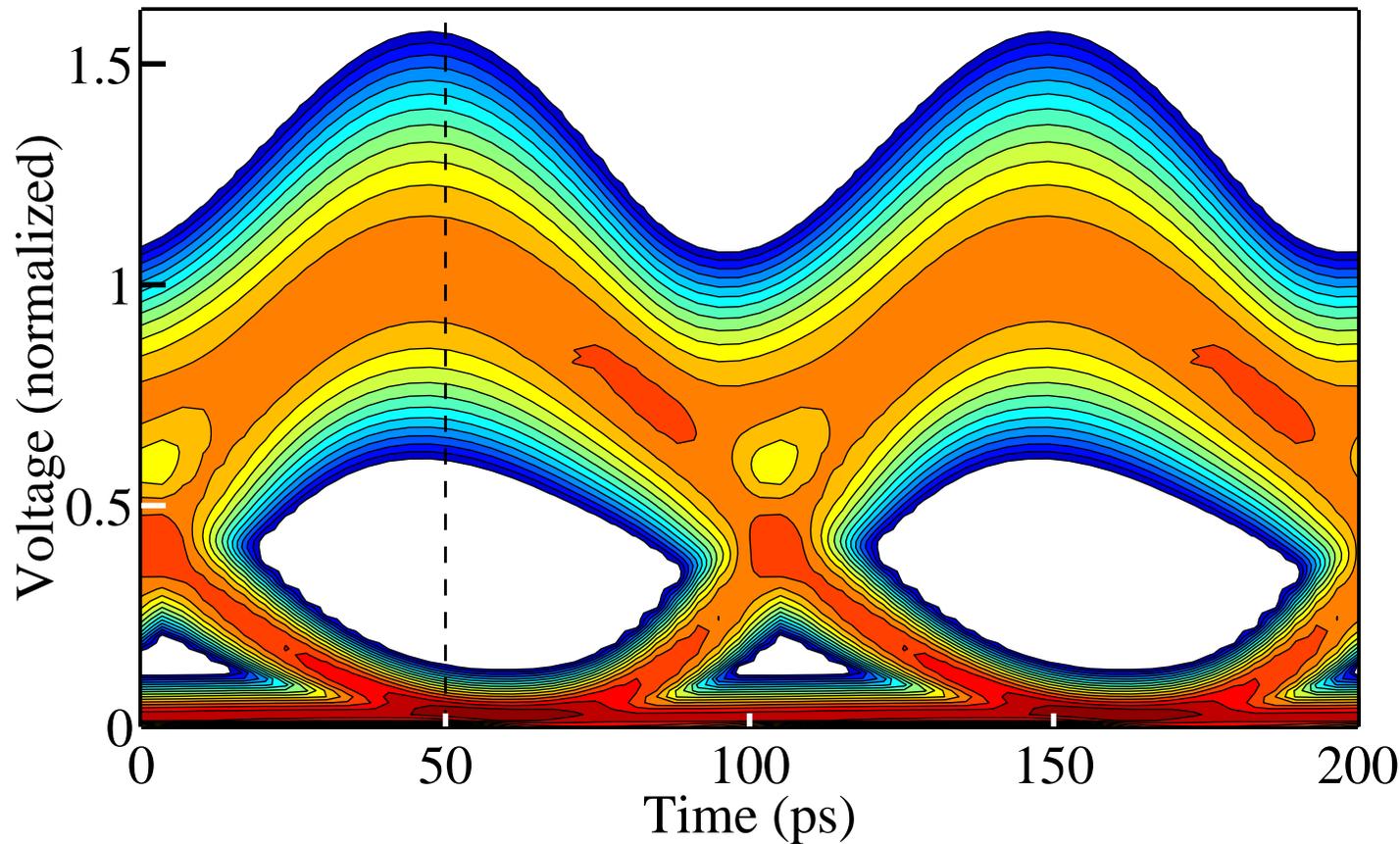


Pdfs of the Electrical Signal





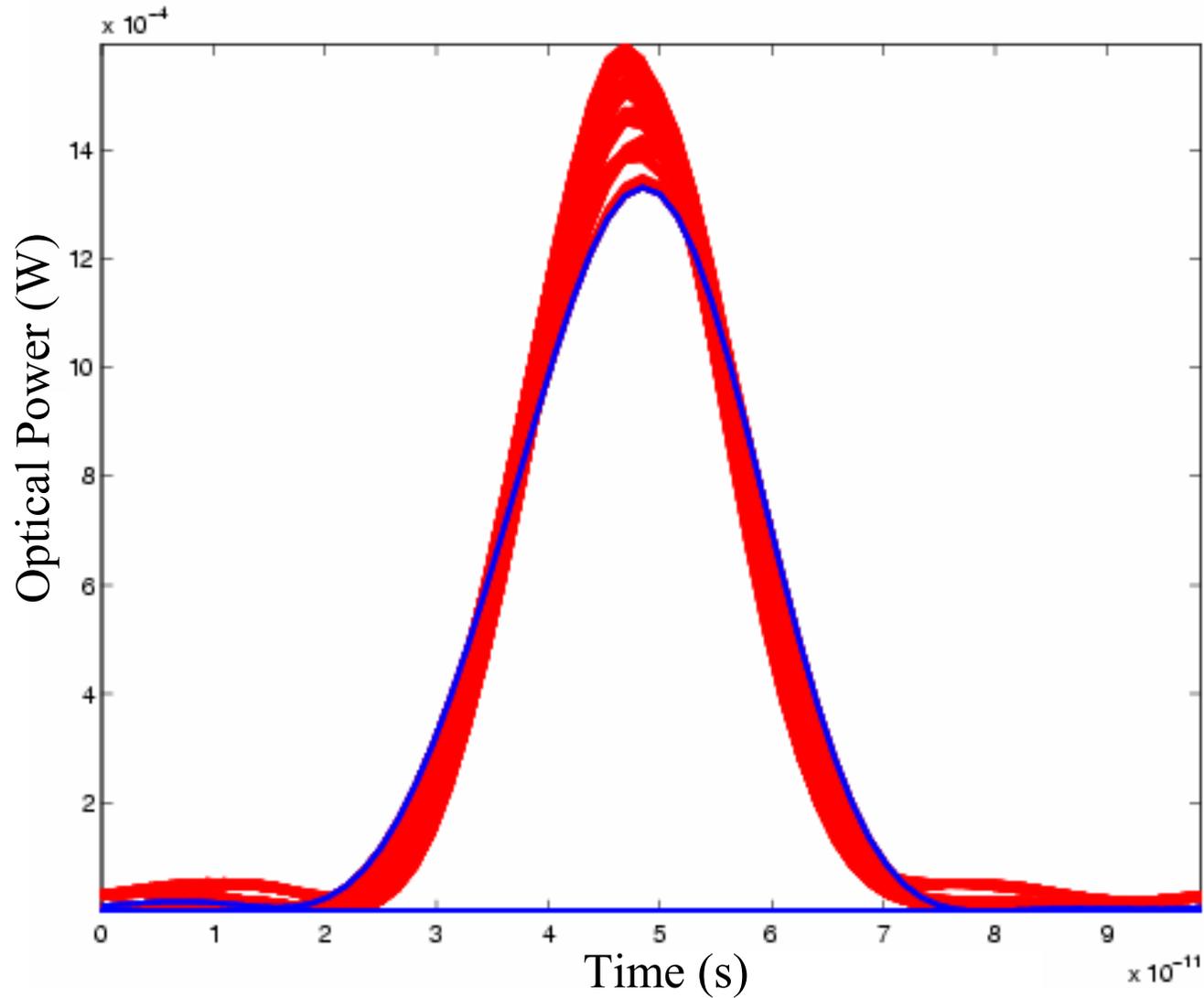
Eye Diagram from Linearization



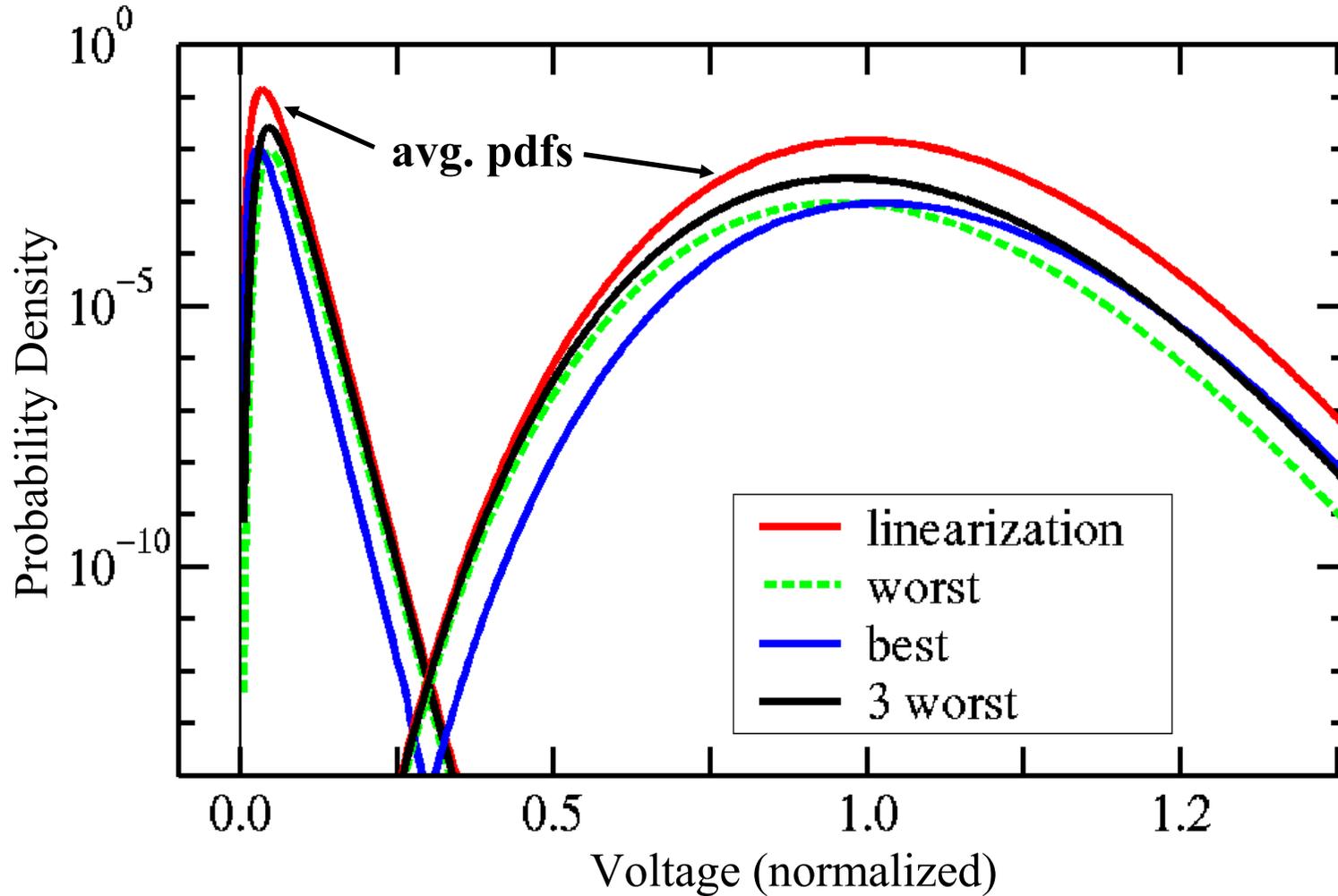
UMBC

Focus on Worst Pattern

by Brian Marks



32 Bits CRZ: Focus on Worst Patterns





Conclusions

- ① Linearization method was successfully applied to CRZ system
- ② Critical steps: Phase jitter separation + dispersion compensation
- ③ Bit patterns are important, focus on worst patterns
- ④ Computational cost equal to $2N$ Monte Carlo noise realizations

Approach might be practical in realistic systems